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A PROGRAMMABLE LOGIC DEVICE HAVING  
AN INTEGRATED PHASE LOCK LOOP

5 Field of the Invention

a The present invention relates generally to programmable logic devices (PLD<sup>S</sup>) and, more particularly to programmable logic devices having an integrated phase lock loop to provide enhanced clocking capabilities and other additional features.

10 Background of the Invention

a Modern computers require various clocks operating at different frequencies to operate different individual components of individual on-board devices. In a programmable logic device (PLD),  
15 to realize various clock frequencies at the particular macro cells (or registers) of the device, previous approaches have traced multiple clock signals throughout the layout of the chip to supply the particular cells with the desired frequencies.

20 Modern semiconductor manufacturers typically specialize in specific component manufacturing processes in which they have expertise. For example, a manufacturer skilled in the fabrication of programmable logic devices may not necessarily be skilled in the manufacturing of phase locked loop (PLL) devices.

Personal Computer (PC) motherboard applications need a standard set of frequencies to operate. These frequencies are typically generated from a reference clock frequency. Since many designs use multiples of certain input frequencies, design engineers typically use delay loops or counters on a PLD to achieve the various frequencies. Consequentially, the logic resources available in the programmable logic device are expended to implement this remedial frequency adjustment. As a result, either less programmable features may be implemented, or either more costly PLD, complex programmable devices (CPLDs) must be implemented or field programmable gate arrays (FPGAs).

Another problem occurs when industry standards change. When standards change, design engineers typically must redesign their entire chips. For example, the peripheral connect interface (PCI) bus currently uses a bus speed of 33 MHz. It is anticipated that the industry standard for the PCI bus will be increased to 66 MHz in the future. The use of previous approaches (such as delay loops in the programming elements of the logic device) would require a significant amount of design work to upgrade to 66 MHz or any other new standard. By reducing setup times, a performance improvement may be realized.

Summary of the Invention

The present invention integrates a phase lock loop (PLL) with a programmable logic device (PLD) to realize a flexible PLD with a variety of clocking options. The present invention generates multiple clock frequencies internally to a programmable logic device using a single reference clock input. The programmer can dynamically change the functionality of the programmable logic device. As a result, a "virtual hardware device" is realized. The ability to change the frequency of operation also dynamically offers a tremendous advantage to users of reconfigurable computing.

Objects, features and advantages of the present invention include providing a dynamically programmable multiple frequency clock generator with a programmable logic device which will create a device more efficient than either of the two devices considered separately. The present invention will provide a wide output frequency range that can be dynamically adjusted, a number of individually programmable outputs, a high degree of control of output skew, an internal loop filtering which would not require external components and a wide number of output frequencies. The present invention may be configured to feed a clock distribution network of targeted programmable logic devices and may be accessible to one or more input/output (I/O) pins. In a particular

embodiment, the present invention may provide a low clock jitter (less than 200 ps), a variable duty cycle (ranging between 40% and 60%), either a 3.3 volt or 5.0 volt input supply voltage operation range, a matched output impedance and a low power consumption. The present invention may be implemented using high speed CMOS implementation.

#### Brief Description of the Drawings

These and other objects, features and advantages of the present invention will be apparent from the following detailed description and the appended claims in which:

FIG. 1 is a block diagram illustrating a preferred embodiment of the present invention incorporated into a CPLD architecture;

FIG. 2 is a block diagram of an individual distribution cell of a preferred embodiment of the present invention;

FIG. 3 is a timing diagram illustrating the falling edge triggered clock of a preferred embodiment of the present invention;

FIG. 4 is a block diagram illustrating a flip-flop scheme for implementing the falling edge triggered flip-flops;

FIG. 5 is a timing diagram illustrating a relationship between a PHI1 and a PHI2 signal;

FIG. ~~8~~ is a flip-flop scheme illustrating the implementation of the PHI1 and PHI2 signals; and

FIG. ~~7~~ is a block diagram illustrating an alternate embodiment of the present invention including a multiplexer for adding additional flexibility.

#### Detailed Description of the Preferred Embodiments

A block diagram of a CPLD 10 incorporating a preferred embodiment of the present invention is shown. The CPLD 10 generally comprises an input section 12, a logic section 14, a logic section 16 and a Programmable Interconnect Matrix (PIM) 18. The input section generally comprises a PLL structure 20. The PLL structure 20 has an input 22 that receives a clock from an external source (not shown). The PLL structure 20 produces a number of individual clocks on a multi-bit bus 24 that is shown, for example, as being a 4-bit bus. A 4-bit bus may produce four individual clock signals that are presented to the PIM 18. A number of programming inputs are received at a multi-bit bus 26. The multi-bit bus 26 presents these inputs to the PIM 18. The multi-bit bus 24 presents the individual clock inputs to the logic section 14 and the logic section 16. A feedback of the clock inputs is also presented back to the multi-bit bus 26.

Referring to FIG. 2, a block diagram of a clock distribution scheme 30 in accordance with a preferred embodiment of the present invention is shown. The clock distribution scheme 30 generally comprises an input 32, an input 34, an output 36, an output 38, an output 40 and an output 42. The input 32 may receive a reference clock frequency from an external device (not shown). The input 34 may receive configuration information from a control logic (not shown). The output 36 generally presents a signal CLK0, the output 38 generally presents a signal CLK1, the output 40 generally presents a signal CLK2 and the output 42 generally presents a signal CLK3. The outputs 36, 38, 40 and 42 may be presented to a clock distribution network on a programmable logic device (not shown). Each of the outputs 36, 38, 40 and 42 may be configured to operate at an independent frequency that may drive the individual logic blocks of the programmable logic device. A detailed illustration of how to perform such an independent clock configuration may be found in U.S. Patent Application No. 08/549,915, which is hereby incorporated by reference in its entirety. While a particular aspect of cited reference deals with using a non-volatile memory such as an EPROM to produce and configure the desired clocks, the present invention may be implemented using a wider variety of PLLs and PLDs.

Referring to FIG. 3, a timing diagram 43 illustrating a falling edge triggered flip-flop is shown. The timing diagram 43 generally comprises a reference clock signal 44 and an output clock signal 46. The reference clock signal 44 is generally a fixed frequency clock that may be generated either internally or externally in order to fit the design criteria of a particular application. The output clock 46 has a single pulse 48 that is skewed from the individual pluses of the reference clock 44 by a fixed amount  $T_{co}$ .

Referring to FIG. 4, a block diagram illustrating the implementation of a falling edge triggered flip-flop scheme 50 is shown. The scheme 50 generally comprises a first flip-flop 51, a second flip-flop 52 and a third flip-flop 53. The flip-flop 51 has an input 56 that may receive a reference clock CLK. The flip-flop 52 has an input 57 that may receive the reference clock CLK. The flip-flop 53 has an input 58 that may receive the clock CLK. Each of the flip-flops 51, 52 and 53 has an input D and an output Q. The flip-flops 51, 52 and 53 are generally cascaded together. The output Q of the flip-flop 53 provides a signal OUT that corresponds to the output signal 46 of FIG. 3. The flip-flops 51, 52 and 53 are generally edge triggered devices.

Referring to FIG. 5, a timing diagram illustrating a relationship between a signal PHI1 and a signal PHI2 is shown. The signal PHI1 is generally a fixed frequency clock. The signal PHI2 is also generally a fixed frequency clock. The signal PHI1 and PHI2 are generally out of phase by a fixed amount Tco. An output signal 59 is triggered at the end of the fixed amount Tco.

Referring to FIG. 6, a flop-flop scheme 60 illustrating the implementation of the signal PHI1 and PHI2 is shown. The flip-flop scheme 60 generally comprises a flip-flop 62, a flip-flop 64 and a flip-flop 66. The flip-flop 62 has an input 67 that generally receives the signal PHI1 and the flip-flop 64 has an input 68 that generally receive the signal PHI1. Similarly, the flip-flop 66 has an input 69 that generally receive the signal PHI2. Each of the flip-flops 62, 64 and 66 have an input D and an output Q. The flip-flops 62, 64 and 66 are generally cascaded together. The output Q of the flip-flop 66 generally provides the output OUT shown in FIG. 5. The flip-flop scheme 60 allows a zero delay input and/or output buffer to be implemented. The zero delay input buffer allows set up (Ts) and hold (Th) times to be adjusted to meet high frequency design requirements. The zero delay output buffer allows adjustment of the clock to an output delay (Tco) to meet the design criteria of a particular application. As a result,



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programmers may run their designs at very high frequencies while eliminating the delays involved with the Tco and Ts times.

Referring to FIG. 7, a block diagram illustrating an alternate embodiment clock distribution scheme 70 of the present invention is shown. The clock distribution scheme 70 generally comprises a multiplexer 72 and a clock distribution block 74. The multiplexer 72 has a first input 76 that generally receives an internally generated clock, an input 78 that generally receives an externally generated clock and an input 80 that generally receives a configuration signal that selects between the first input 76 and the second input 78. The multiplexer 72 presents a clock signal at the output 82 that is received at an input 84 of the clock distribution block 74. The clock distribution block 74 generally comprises an output 86, an output 88, an output 90, an output 92 and a control output 94. The output 86 generally presents a signal CLK0, the output 88 generally presents a signal CLK1, the output 90 generally presents a signal CLK2 and the output 92 generally presents a signal CLK3. The outputs 86, 88, 90 and 92 may be presented to a clock distribution network on a programmable logic device (not shown). Each of the outputs 86, 88, 90 and 92 may be configured to operate an independent frequency that may drive the individual logic blocks of the programmable logic device. Each of

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the clock signals CLK0, CLK1, CLK2 and CLK3 are individually programmable to a plurality of frequencies. The clock distribution block 74 may provide the individually programmable frequencies at the outputs 86, 88, 90 and 92 by any of a plurality of means including, but not limited to, a phase lock loop (PLL). Each of the signals CLK0, CLK1, CLK2 and CLK3 are accessible through one or more input/output pins. Additionally, each of the outputs 86, 88, 90 and 92 may have a particular output impedance that may be adjusted to match the impedance of an external device. Since the frequencies present at the outputs 86, 88, 90 and 92 are controlled in part by the control signal received at the input 94, the frequencies may be programmed after fabrication of the clock distribution scheme 70.

The input 76 of the multiplexer 72 may receive one or more internally generated clocks. Similarly, the input 78 to the multiplexer 72 may receive one or more externally generated clocks. As a result, the multiplexer 72 may provide a plurality of reference clocks at the input 84. Since a plurality of reference clocks may be present at the input 84, the manipulation provided by the clock distribution block 74 is enhanced to provide even a greater number of frequencies at the outputs 86, 88, 90 and 92. The clock distribution scheme 70 may be implemented in a

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programmable logic device or a complex programmable logic device according to the design criteria of a particular application. The number of clocks present at the input 76 may be adjusted to fit the design criteria of a particular application. The number of configuration bits present at the input 94 may be adjusted to fit the design criteria of a particular application.

The present invention integrates a PLL with a PLD to realize a flexible PLD with a variety of clocking options. The present invention generates multiple clock frequencies internally to a programmable logic device using a single reference clock input. The present invention may also be implemented using a field programmable gate array (FPGA).

While the invention has been particularly shown and described with reference to preferred embodiments thereof, it will be understood by those skilled in the art that various changes in form and details may be made without departing from the spirit and scope of the invention.